

# The RF Sub-Micron Bipolar Line

## RF Power Bipolar Transistors

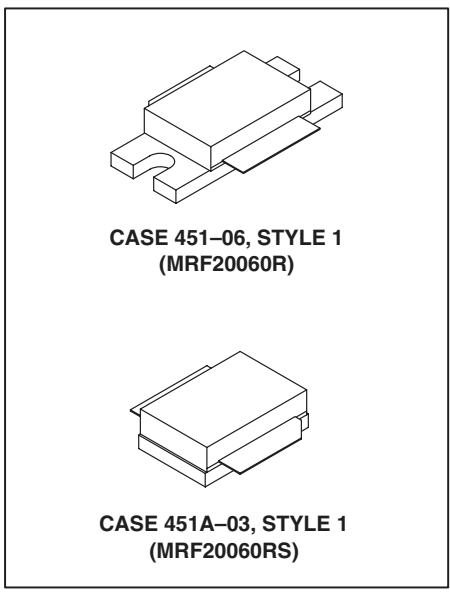
The MRF20060R and MRF20060RS are designed for class AB broadband commercial and industrial applications at frequencies from 1800 to 2000 MHz. The high gain, excellent linearity and broadband performance of these devices make them ideal for large-signal, common emitter class AB amplifier applications. These devices are suitable for frequency modulated, amplitude modulated and multi-carrier base station RF power amplifiers.

- Guaranteed Two-tone Performance at 2000 MHz, 26 Volts  
 Output Power — 60 Watts (PEP)  
 Power Gain — 9 dB  
 Efficiency — 33%  
 Intermodulation Distortion — -30 dBc
- Characterized with Series Equivalent Large-Signal Impedance Parameters
- S-Parameter Characterization at High Bias Levels
- Excellent Thermal Stability
- Capable of Handling 3:1 VSWR @ 26 Vdc, 2000 MHz, 60 Watts (PEP) Output Power
- Designed for FM, TDMA, CDMA and Multi-Carrier Applications
- Test Fixtures Available at: <http://mot-sps.com/rf/designtds/>

**Note:** Not suitable for class A operation.

**MRF20060R**  
**MRF20060RS**

**60 W, 2000 MHz**  
**RF POWER**  
**BROADBAND**  
**NPN BIPOLAR**



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### MAXIMUM RATINGS

Rating	Symbol	Value	Unit
Collector-Emitter Voltage ( $I_B = 0$ mA)	$V_{CEO}$	25	Vdc
Collector-Emitter Voltage	$V_{CES}$	60	Vdc
Collector-Base Voltage	$V_{CBO}$	60	Vdc
Collector-Emitter Voltage ( $R_{BE} = 100$ Ohm)	$V_{CER}$	30	Vdc
Base-Emitter Voltage	$V_{EB}$	- 3	Vdc
Collector Current - Continuous	$I_C$	8	Adc
Total Device Dissipation @ $T_C = 25^\circ\text{C}$ Derate above $25^\circ\text{C}$	$P_D$	250 1.43	Watts W/ $^\circ\text{C}$
Storage Temperature Range	$T_{stg}$	- 65 to +150	$^\circ\text{C}$
Operating Junction Temperature	$T_J$	200	$^\circ\text{C}$

### THERMAL CHARACTERISTICS

Rating	Symbol	Max	Unit
Thermal Resistance, Junction to Case	$R_{\theta JC}$	0.7	$^\circ\text{C}/\text{W}$

**ELECTRICAL CHARACTERISTICS** ( $T_C = 25^\circ\text{C}$  unless otherwise noted)

Characteristic	Symbol	Min	Typ	Max	Unit
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**OFF CHARACTERISTICS**

Collector–Emitter Breakdown Voltage ( $I_C = 50\text{ mAdc}$ , $I_B = 0$ )	$V_{(BR)CEO}$	25	28	—	Vdc
Collector–Emitter Breakdown Voltage ( $I_C = 50\text{ mAdc}$ , $V_{BE} = 0$ )	$V_{(BR)CES}$	60	69	—	Vdc
Collector–Base Breakdown Voltage ( $I_C = 50\text{ mAdc}$ , $I_E = 0$ )	$V_{(BR)CBO}$	60	69	—	Vdc
Reverse Base–Emitter Breakdown Voltage ( $I_B = 10\text{ mAdc}$ , $I_C = 0$ )	$V_{(BR)EBO}$	3	3.5	—	Vdc
Zero Base Voltage Collector Leakage Current ( $V_{CE} = 30\text{ Vdc}$ , $V_{BE} = 0$ )	$I_{CES}$	—	—	10	mAdc

**ON CHARACTERISTICS**

DC Current Gain ( $V_{CE} = 5\text{ Vdc}$ , $I_C = 1\text{ Adc}$ )	$h_{FE}$	20	40	80	—
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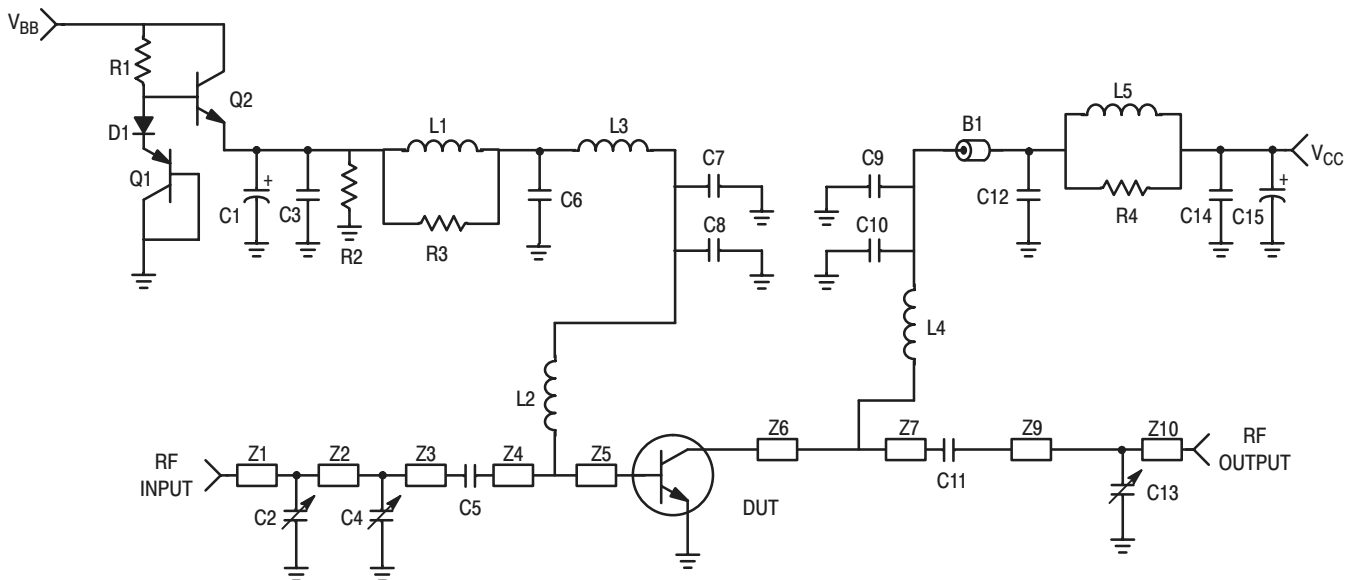
**DYNAMIC CHARACTERISTICS**

Output Capacitance ( $V_{CB} = 26\text{ Vdc}$ , $I_E = 0$ , $f = 1.0\text{ MHz}$ ) <sup>(1)</sup>	$C_{ob}$	—	55	—	pF
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**FUNCTIONAL TESTS** (In Motorola Test Fixture)

Common–Emitter Amplifier Power Gain ( $V_{CC} = 26\text{ Vdc}$ , $P_{out} = 60\text{ Watts (PEP)}$ , $I_{CQ} = 200\text{ mA}$ , $f_1 = 2000.0\text{ MHz}$ , $f_2 = 2000.1\text{ MHz}$ )	$G_{pe}$	9	9.8	—	dB
Collector Efficiency ( $V_{CC} = 26\text{ Vdc}$ , $P_{out} = 60\text{ Watts (PEP)}$ , $I_{CQ} = 200\text{ mA}$ , $f_1 = 2000.0\text{ MHz}$ , $f_2 = 2000.1\text{ MHz}$ )	$\eta$	33	35	—	%
Intermodulation Distortion ( $V_{CC} = 26\text{ Vdc}$ , $P_{out} = 60\text{ Watts (PEP)}$ , $I_{CQ} = 200\text{ mA}$ , $f_1 = 2000.0\text{ MHz}$ , $f_2 = 2000.1\text{ MHz}$ )	IMD	—	–32	–30	dB
Input Return Loss ( $V_{CC} = 26\text{ Vdc}$ , $P_{out} = 60\text{ Watts (PEP)}$ , $I_{CQ} = 200\text{ mA}$ , $f_1 = 2000.0\text{ MHz}$ , $f_2 = 2000.1\text{ MHz}$ )	IRL	12	19	—	dB
Output Mismatch Stress ( $V_{CC} = 26\text{ Vdc}$ , $P_{out} = 60\text{ Watts (PEP)}$ , $I_{CQ} = 200\text{ mA}$ , $f_1 = 2000.0\text{ MHz}$ , $f_2 = 2000.1\text{ MHz}$ , VSWR = 3:1, All Phase Angles at Frequency of Test)	$\psi$	No Degradation in Output Power			

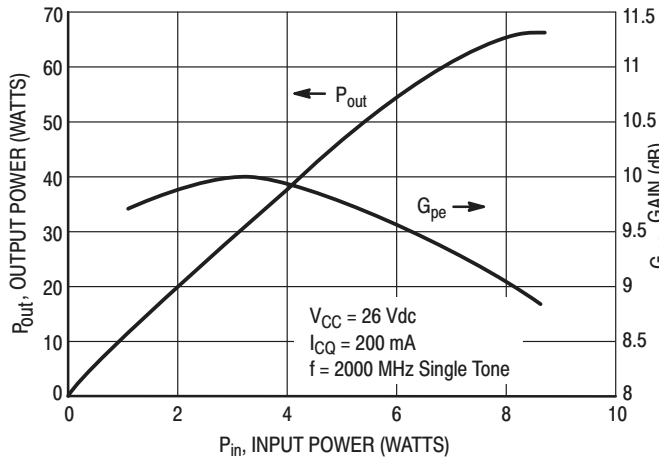
(1) For Information Only. This Part Is Collector Matched.



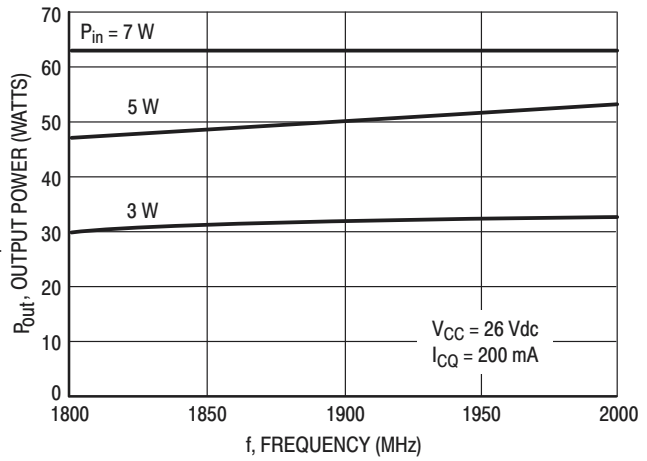
B1	Ferrite Bead, P/N 5659065/3B, Ferroxcube	D1	Diode, Motorola (MURS160T3)
C1	100 $\mu$ F, 50 V, Electrolytic Capacitor, Mallory	L1, L5	12 Turns, 22 AWG, 0.140" Choke
C2, C4, C13	0.6–4.0 pF, Variable Capacitor, Gigatrim, Johanson	L2, L4	.5 inch of 20 AWG
C3, C14	0.1 $\mu$ F, Chip Capacitor, Kemit	L3	12.5 nH Inductor
C5	15 pF, B Case Chip Capacitor, ATC	R1	2 x 130 $\Omega$ , 1/8 W Chip Resistor, Rohm
C6, C12	1000 pF, B Case Chip Capacitor, ATC	R2	2 x 100 $\Omega$ , 1/8 W Chip Resistor, Rohm
C7, C9	91 pF, B Case Chip Capacitor, ATC	R3, R4	10 $\Omega$ , 1/2 W, Resistor
C8, C10	24 pF, B Case Chip Capacitor, ATC	Q1	Transistor, PNP Motorola (BD136)
C11	13 pF, B Case Chip Capacitor, ATC	Q2	Transistor, NPN Motorola (MJD47)
C15	470 $\mu$ F, 50 V, Electrolytic Capacitor, Mallory	Board	Glass Teflon <sup>®</sup> , Arlon GX-0300-55-22, $\epsilon_r$

Figure 1. 1.93 – 2 GHz Test Fixture Electrical Schematic

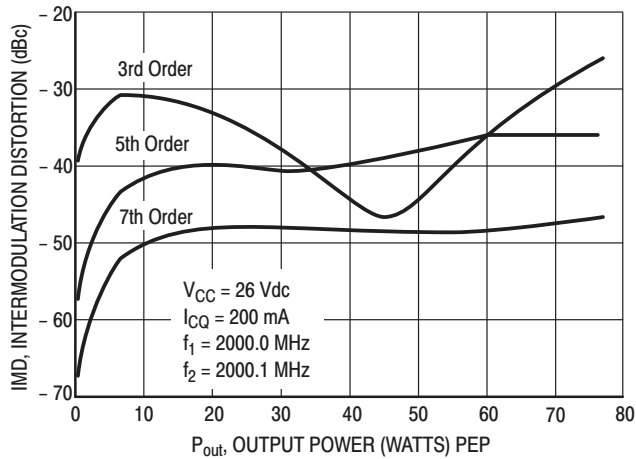
### TYPICAL CHARACTERISTICS



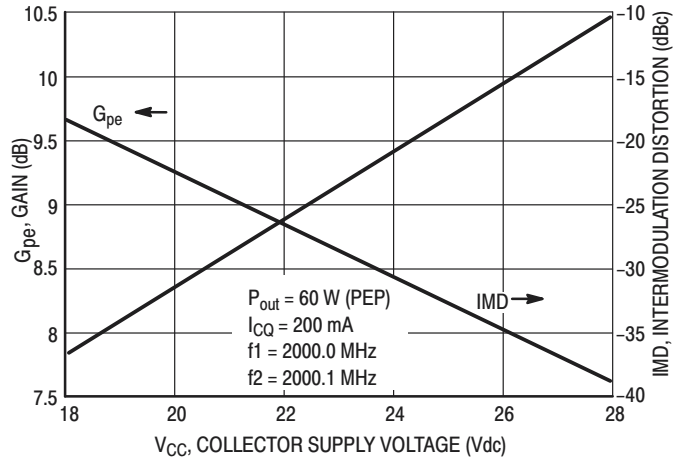
**Figure 2. Output Power & Power Gain versus Input Power**



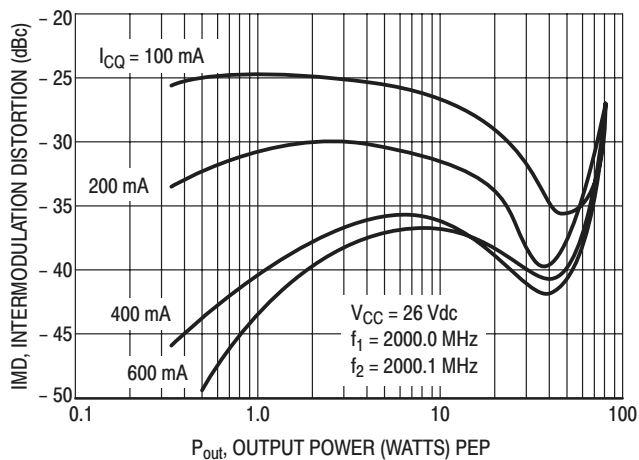
**Figure 3. Output Power versus Frequency**



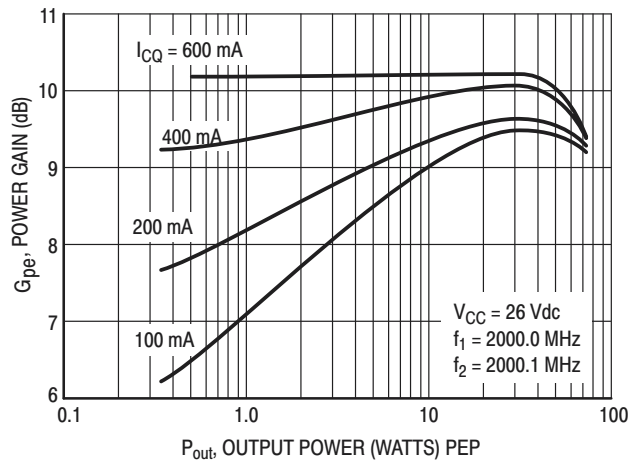
**Figure 4. Intermodulation Distortion versus Output Power**



**Figure 5. Power Gain and Intermodulation Distortion versus Supply Voltage**



**Figure 6. Intermodulation Distortion versus Output Power**



**Figure 7. Power Gain versus Output Power**

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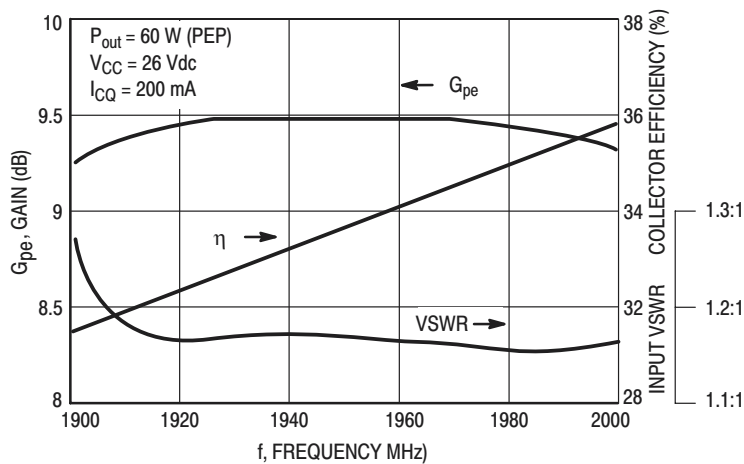
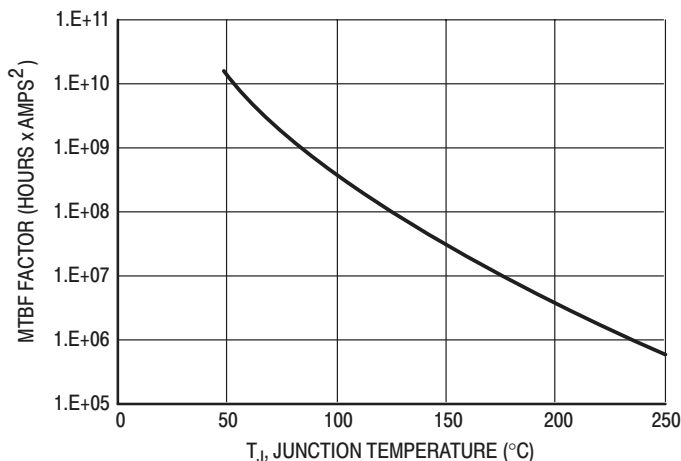
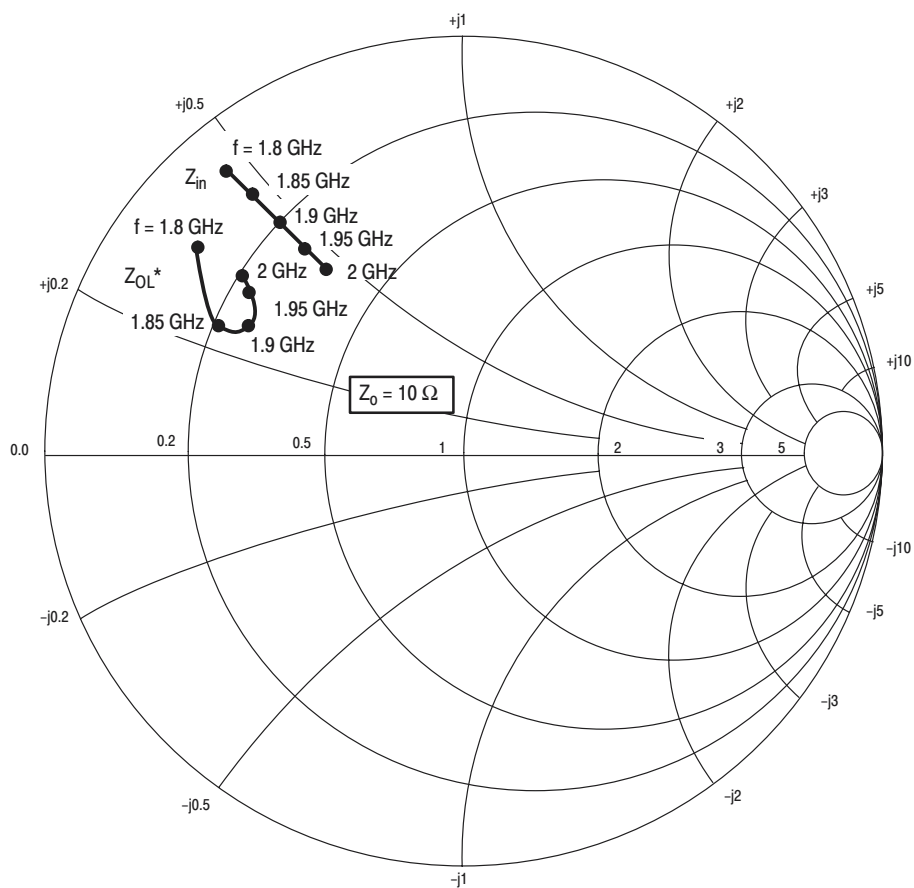


Figure 8. Performance in Broadband Circuit



This above graph displays calculated MTBF in hours x ampere<sup>2</sup> emitter current. Life tests at elevated temperatures have correlated to better than  $\pm 10\%$  of the theoretical prediction for metal failure. Divide MTBF factor by  $I_C^2$  for MTBF in a particular application.

Figure 9. MTBF Factor versus Junction Temperature



$V_{CC} = 26 \text{ V}$ ,  $I_{CQ} = 200 \text{ mA}$ ,  $P_{out} = 60 \text{ W (PEP)}$

f MHz	$Z_{in}(1)$ $\Omega$	$Z_{OL}^*$ $\Omega$
1800	$1.0 + j4.8$	$1.7 + j3.3$
1850	$1.5 + j4.8$	$2.2 + j2.7$
1900	$2.0 + j4.7$	$2.4 + j3.0$
1950	$2.5 + j4.7$	$2.3 + j3.2$
2000	$3.5 + j4.7$	$2.0 + j3.4$

$Z_{in}(1)$  = Conjugate of fixture base terminal impedance.

$Z_{OL}^*$  = Conjugate of the optimum load impedance at given output power, voltage, bias current and frequency.

**Figure 10. Series Equivalent Input and Output Impedance**

Table 1. Common Emitter S-Parameters at  $V_{CE} = 24 \text{ Vdc}$ ,  $I_C = 3.5 \text{ Adc}$

f GHz	S <sub>11</sub>		S <sub>21</sub>		S <sub>12</sub>		S <sub>22</sub>	
	S <sub>11</sub>	∠ φ	S <sub>21</sub>	∠ φ	S <sub>12</sub>	∠ φ	S <sub>22</sub>	∠ φ
1.5	0.986	168	0.32	81	0.031	60	0.923	169
1.55	0.985	167	0.35	76	0.031	63	0.918	169
1.6	0.981	167	0.40	70	0.032	61	0.908	169
1.65	0.973	166	0.45	63	0.030	53	0.897	169
1.7	0.968	165	0.52	56	0.033	50	0.889	168
1.75	0.951	163	0.62	46	0.028	47	0.880	169
1.8	0.914	161	0.76	32	0.027	39	0.871	170
1.85	0.851	161	0.91	12	0.024	26	0.863	171
1.9	0.789	164	1.02	-15	0.015	5	0.888	174
1.95	0.810	170	0.94	-44	0.005	-7	0.931	174
2	0.880	172	0.75	-68	0.006	-151	0.953	172
2.05	0.934	170	0.57	-85	0.010	152	0.967	170
2.1	0.964	168	0.45	-98	0.015	158	0.965	169
2.15	0.977	165	0.36	-109	0.022	164	0.955	168
2.2	0.975	163	0.30	-118	0.033	165	0.950	167
2.25	0.961	161	0.25	-128	0.049	160	0.947	167
2.3	0.942	160	0.22	-139	0.066	149	0.938	166
2.35	0.919	157	0.19	-149	0.077	142	0.931	165
2.4	0.860	156	0.17	-163	0.100	137	0.922	165
2.45	0.821	159	0.15	177	0.128	122	0.914	165
2.5	0.781	161	0.14	157.0	0.156	108	0.907	165

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# NOTES



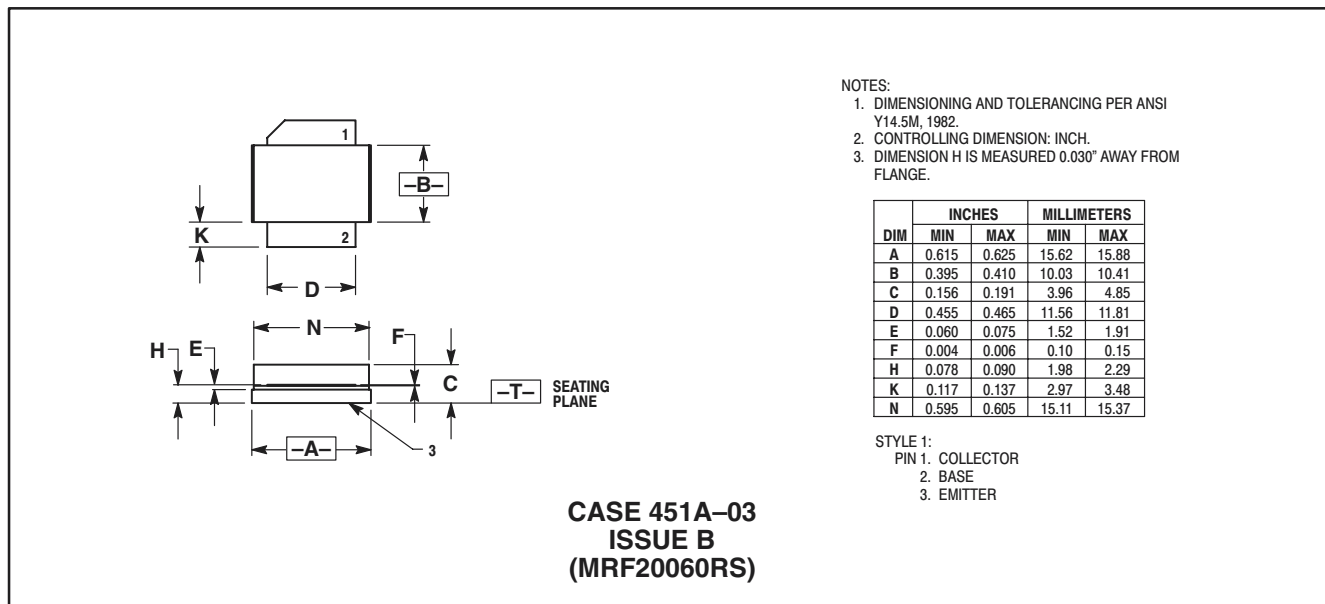
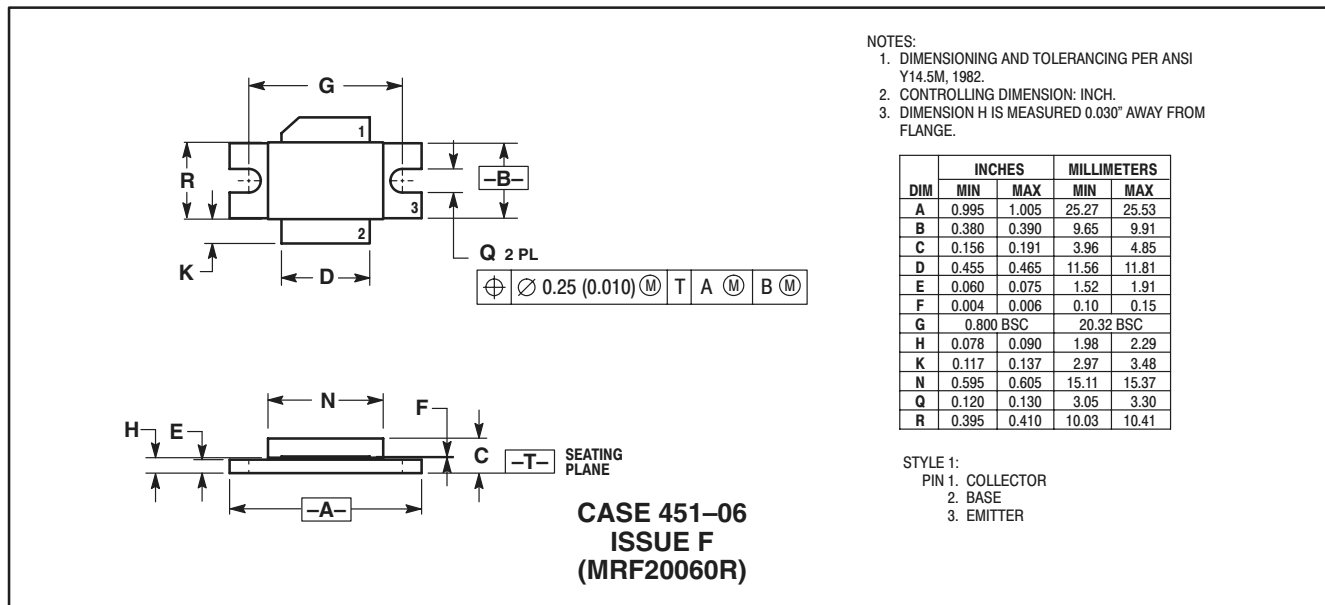


# NOTES




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## PACKAGE DIMENSIONS





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2, Dai King Street, Tai Po Industrial Estate, Tai Po, N.T., Hong Kong.  
852-26668334

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